TECHNICAL SERVICE MANUAL

FOR

HF SSB TRANSCEIVER TYPE 7727

To be used in conjunction with Technical Service Manual for HF SSB Transceiver Type 7515.

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Always quote the handbook number in correspondence on technical matters concerning the maintenance of a unit.

On receipt of amendments, please insert them promptly.

Issue 1 : December 1977.

HF SSB TRANSCEIVER TYPE 7727

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3. BRIEF DESCRIPTION Refer 7515 TSM.

The 7727 does not operate in the AM mode. References to AM operation do not apply.

4. SWITCHING Refer 7515 TSM.

The 7727 does not operate in the AM mode. References to AM operation do not apply.

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Component Layouts and Assembly Drawings :-

PA Assy	08-01290
PA Filter PCB Assy 1	08-01179
PA Filter PCB Assy 2	08-01180

1. GENERAL INFORMATION

1.1 Introduction

The Type 7727 transceiver is similar to the Type 7515, but with an increased power output of 100 W PEP. The differences apply primarily to the Power Amplifier and filter assemblies with minor changes to the remainder of the equipment.

This supplement covers only those areas where a difference exists and should be used in conjunction with the Type 7515 Technical Service Manual.

1.2 Specifications

1.2.1 General

Design figures are quoted with limit figures in brackets, all tests are carried out at 13.6 V DC input.

Frequency range :	2 – 11 MHz
Crystal capacity :	6
Channel capacity :	6 single frequency simplex, or 3 two frequency simplex, or any combination that does not exceed the crystal capacity.
Operating modes :	Single sideband (A3J) USB, LSB or both (1978)
Frequency stability :	\pm 50 Hz, in range 5°C to 55°C \pm 60 Hz, in range 0°C to 60°C with \pm 10% input voltage variation
Controls :	Power ON/OFF and VOLUME. Mode: USB, LSB, TUNE. CHANNEL SELECT. CLARIFIER.

EMERGENCY CALL (RFDS) NOISE LIMITER ON/OFF

Indicators (illuminated) :

Power Transmit

MUTE ON/OFF

1.2.1 (Cont.)

		Side Rear	Microphone Antenna Power input Extension L.S. (Optional)			
	RF input/output	impedance :	50 ohms nominal. The transceiver will operate with a load impedance producing a VSWR of less than 2:1.			
	Input voltage :		Nominal 13.6 V DC。 The transceiver will operate with input voltages between 10 V and 15V positive or negative ground, reverse polarity protec- tion is provided			15V
	Power consumpt	ion :	Receive, 270mA no signal Transmit, A3J average 5A A3J two-tone 11A (nominal)			
	Environmental :	:	Ambient Temper	rature ^o C	-10 to +30	+30 to +60
			Relative Humidi	ity %	above 95	from 95 at 30 [°] C to max。50 at 60 [°] C
			Atmospheric Pre	essure	sea level) v temperature	rs (3600 m above vith maximum e derated by 1°C 30 m above sea
	Cooling :		Convection			
	Size :		Width : Height : Depth :	27 cm (10 9 cm (3 33 cm (13	.5")	
	Weight :		4.4 kg (9.7 lbs)		
	Receiver					
	Туре:		All transistor, s	single con	version super	heterodyne.
,	IF:		1650kHz			

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1.2.2

1.2.2 (Cont.)

An aerial EMF of less than 0.5uV from a 50 ohm Sensitivity : source will produce a signal plus noise/noise ratio of 10dB (A3J). Maximum input : 10V rms. Selectivity : A3J, better than 70dB (60) -1, +5kHzbetter than 6dB (8), 300 - 2800Hz. Better than 60dB (55). Image Rejection : **Spurious Signal Rejection :** Better than 65dB (60). **Cross Modulation :** A signal 75dB (65), above a signal producing a 10dB S+ N/N ratio modulated 30% and removed at least 20kHz from the wanted signal will produce an increase in receiver noise of less than 3dB. Intermodulation : To produce a third order intermodulation product equivalent to a wanted signal producing a 10dB S + N/N ratio, two unwanted signals greater than 30kHz removed from the wanted signal must have a level greater than 65dB (60) above the wanted signal. Blocking : A signal 85dB (70) above a signal producing a 10dB S + N/N ratio and removed at least 20kHz from the wanted signal will cause a change in output level of the wanted signal of less than 3dB. AGC : Less than 3dB (6) variation in output for signal strength variations between 3 microvolts and 100 millivolts. AF Power and Distortion : 2.5 Watts at less than 5% THD. $300 - 2800 \text{Hz}, -3^{40} \text{dB}$ **AF Response :**

1.2.2 (Cont.)

Clarifier Range

CAN :	Nominal ±0.0007% of SCF or ±25Hz whicheve is greater, continuously variable.	
CAW : **	Nominal ±0.003% of SCF, continuously variable.	

** Export Only

These parameters are measured in terms of Australian P & T Department Specification RB209, RB209-0.

Transmitter

Type :

Power Output :

AF Response :

Harmonic Emissions :

Spurious Emissions :

Carrier Supression :

Unwanted Sideband :

Intermodulation Products :

ALC Range :

Clarifier Range :

Microphone :

All solid state including power stages.

SSB (A3J) : 100 Watts PEP \pm 1dB.

Overall response of transmitter rises at approximately 6dB/octave from 300Hz to 2800Hz. At higher frequencies cut-off is very rapid.

55dB (45) below PEP.

Spurious emissions (not harmonics) separated from the carrier by more than 20kHz, 50dB (43) below PEP.

50dB (40) below PEP.

45dB (43) below PEP.

33dB (31) below PEP.

30dB change in input signal produces less than 1dB change in output.

As for Receiver Specifications above NOTE: Only available when Option CB fitted.

Rocking armature, with push-to-talk fitted to case.

1.3 List of Options and Accessories

100

100 C

- (a) <u>Options</u> Same as 7515 except option A(AM) is NOT available.
- (b) Accessories Same as 7515 except Code 501, AC Power Supply TYPE 7113 must be used, with the appropriate interconnecting cable.

5. TECHNICAL DESCRIPTION

Reference Dwg B04-00822

5.1 Introduction

All components on printed circuit boards are prefixed by a number.

The code is as follows :-

Prefix	PCB Number	Description
1	Annual Control of the	Frame assembly
2	07-00337	Main PCB Assy
3	07-00401	PA Assy
4	07-00409	4.6 to 11 MHz PA filters and Broadcast filter
6	07-00305	Emergency Call assy
7	07-00410	2 – 4.6 MHz PA filters

Example R47 on PCB 07-00401 becomes 3R47.

For clarity of reading the prefix has been omitted in the technical description when referring to components on a board under discussion. If reference is made to a component on another board, the prefix is added to the latter. The order of stages described follows the signal path.

5.2

5.2.1 to Refer 7515 TSM.

5.2.10

5.2.11 Gain Controlled Amplifier and Input Stage

The DC supply is derived from the A rail via transistor switch TR9, TR10 which is controlled by the D rail.

R2 reduces the supply voltage to TR1 to approximately 9V.

The leakage inductance of T1 together with C1, C2 and C3 form a low pass filter to further attenuate any unwanted high frequency signals from the exciter. The gain of TR1 is controlled to maintain the required PEP output by the ALC processor via emitter follower TR2 to pin 7. The output from pin 3 is fed via R4 and frequency compensating network R5, C7 to the bases of TR3 and TR4. Negative feedback is applied around TR3 and TR4 by R7.

5.2.12 PA Driver

The PA driver consists of TR5, TR6 operating in push-pull class B. The bias voltage is provided by TR12 connected as a diode. The bias voltage (and hence the driver standing current) may be increased or reduced by the SOT resistor R44.

The network consisting of R54, C14 and L3 provides high frequency loading for TR5 and TR6.

5.2.13 PA

The DC supply to the PA is taken from the unregulated supply rail via a fuse. Supply is fed via T5 which provides the centre tap for the output transformer T7. TR7 and TR8 form a transformer coupled push-pull amplifier biased by a bias regulator via L5 and T3 to operate in Class B. Negative feedback is provided by T4, R16 and R17.

The bias supply consists of the feedback amplifier TR14 and <u>Darlington</u> transistor TR13. If the bias voltage tends to rise, TR14 passes more current and its collector voltage falls reducing the output voltage of TR15 and hence reducing the bias voltage.

TR14 is thermally bonded to the PA transistor heat sink and provides temperature compensation for the output transistors. Zener diode Z3 stabilises the current in TR14 against power supply variations. Fine adjustment of the bias voltage is provided by R51.

The DC component of the drive current (i.e. the collector current in TR13) passes through the TRANSMIT lamp and hence the lamp brilliance varies with power output. Zener diode Z4 protects the lamp from over voltage when the bias current exceeds the lamp rating. R53 may be fitted to increase or decrease the lamp sensitivity, as the bias current depends on the current gain of the PA transistors.

5.2.14 PA Filter Assemblies

Four separate low pass filters are provided to cover the frequency range of the transceiver and are selected by SW1 h & e.

5.2.15 ALC Detectors

A composite ALC signal is derived from three detectors.

- (a) Voltage Detector The voltage at the collectors of TR7 and TR8 is sampled by T6 and rectified by D2, R9 and C21 providing frequency shaping.
- (b) Forward Power Detector A voltage proportional to the forward power output is developed across T10 and the appropriate secondary of T9 and is rectified by D4.
- (c) <u>Reflected Power Detector</u> A voltage proportional to the reflected power is developed across T8 and the appropriate secondary of T9 and is rectified by D3

The three ALC voltages are OR-ed and used to generate the ALC voltage for the Gain Controlled Amplifier.

Under normal load conditions (50 ohms) the forward power detector produces the largest output and operates the ALC. When a mismatched load is applied which increases the PA collector swing the voltage detector output increases and takes control of the ALC, maintaining a constant voltage swing at the PA collector. If the load mismatch does not increase the PA collector voltage swing, the reflected power detector will take control of the ALC when the reflected power exceeds 5W (i.e. approximately 1.6:1 VSWR) and maintain the reflected power constant with increasing VSWR.

5.2.16 ALC Processor

A reference voltage, which is proportional to the supply voltage when the supply voltage is low, and constant when the supply voltage is high, is generated by R42, R43 and Z2. This voltage is fed to the inverting input of TR11b via R39, R37 and R35.

The ALC detector "peak" voltage is divided by R41 and R38 and applied to D8. The detector voltage is also fed to D7 via the "averaging" circuit R40, and C25. The "peak" and "average" voltages are OR-ed by D7 and D8 and applied to the non-inverting input of TR11b. Negative feedback is applied around TR11b by R33 to set its gain.

The output of TR11b is fed via D5 to emitter follower TR2 which in turn controls the gain of TR1. Diode D5 generates a fast attack and slow decay characteristic. R39 is selected to set the power output required.

AMENDMENT No. 7, Maintenance Handbook for Transceivers Type 7515/7727 Sections

The following paragraphs should be substituted for those of the same number for use with 7515/7727 10-Channel series transceivers.

6.2 Channel Set Up Procedure 7515/7727 10-Channel

6.2.1 Components

The table below itemises all the components required to add one frequency to a 7515/7727/7727-T 10 Channel series transceiver. All the components are located on the Main P.C.B..

Table 6.'	
-----------	--

Circuit Annotation	Component type	Qty (per freq)
2D 6-17, 20-31, 36-41, 73-92 2R 47-52, 166-169 2C 37-42, 122-125 2L 2-7, 11-16, 19-26 2C 8-13, 24-29, 126-133 2X 2-11 2C 134-143 2C 134a-143a	Diode 23-10001 Resistor 4700 5% 1/3W Capacitor 47nF 63V Ceramic RF Coil 44-70031 Tuning Capacitor (ref table 6.3) Crystal (ref section 6.4) Compensation Capacitors (ref Section 6.4) Trimmer 2.5pF (e.g. Philips type 808)	5 1 2 2 1 1-3 1

6.2.1 <u>Two Frequency Simplex</u>

The transceiver has the capacity for ten crystals, all of which may be used for single frequency and two frequency simplex applications. For the two frequency simplex channels, two sets of components as per table 6.1 are required plus :-

(i) 1R3 & IR4, both 22 ohms (ii) Switch 14–00133

if a two frequency simplex channel is not already fitted.

6.2.2 Channel Crystal Assy.

Delete this paragraph.

6.2.4 Channel Switch Modifications

Delete this paragraph.

6. SET-UP PROCEDURE

- 6.1 Introduction Refer 7515 TSM
- 6.2 Channel Set-up Procedure Refer 7515 TSM

with the following changes :-

6.2.1 Two-frequency simplex

All six crystals may be used for single frequency simplex channels.

6.2.2 to Refer 7515 TSM

6.2.6

6.2.7

- (e) A power supply which can be set at 13.6 V ± 0.2V is required.
 A Type 7113 PSU may be used.
- 6.2.8 (a) Set the power supply to $13.6 \vee \pm 0.2 \vee$.
- 6.2.9 Refer 7515 TSM
- 6.2.10 (e) Switch to "TUNE" and operate the PTT button.

(h) The output voltage should be approximately 150 V pp.

6.3 Pre-set Adjustments

6.3.1

to Refer 7515 TSM

6.3.5

6.3.6 Carrier Reinsertion

Although AM is not available in the 7727 the carrier level must still be set. (Used in the TUNE mode, for adjustment of antennas and antenna tuners.)

(a) Remove SK1 and monitor the exciter output (SK1/1 and 3) with the CRO. With the transceiver in the SSB transmit mode, apply a 1 kHz tone via the text unit and set the level so that the Microphone Amplifier is in compression by approximately 10dB. Note the level at SK1.

- 6.3.6 (Cont.)
 - (b) Remove the audio drive and switch to the TUNE mode. Adjust 2C82 "CARRIER LEVEL" until the output at SK1 is between 50% and 75% of the level noted in part (a).
- 6.3.7 Refer 7515 TSM
- 6.3.8 Driver and PA Bias

Ensure SK1 is disconnected from the MAIN PCB.

(a) PA Bias

Remove the PA FUSE link and insert a DC ammeter set to the 1A range. In the SSB mode, operate the PTT switch and observe the ammeter reading. Select a value for 3R51 such that the standing current is between 450 and 550 mA.

(b) Driver Bias

Remove the driver stage LINK and insert a DC ammeter set to the 100mA range. In the SSB mode, operate the PTT switch and observe the ammeter reading. Select a value for 3R44 such that the driver standing current is between 20 and 25 mA.

6.3.9 ALC

Replace SK1 and ensure that the RF dummy load is connected to the antenna socket. Connect the CRO and the distortion meter (6918A) RF input (through 47k ohms) across the dummy load. Connect the audio output of the distortion meter to the test unit (AF in). Select the lowest frequency channel on the transceiver. Switch to SSB and operate the PTT switch. Adjust the value of 3R39 to set the two-tone SSB output to 100 W PEP (i.e. 200 V pp). Check the IMD (Intermodulation distortion) on all channels.

The power output may be decreased if necessary to give -27dB IMD on the worst channel.

- N.B. -27dB IMD on the HP6918A is relative to either tone and is equivalent to -33dB IMD with respect to PEP.
- 6.3.10 Refer 7515 TSM
- 6.3.11 (d) The power output on all channels should be $100 \text{ W PEP} \pm 1 \text{ dB}$.
- 6.3.12 (f) Mute Adjustment for AM does not apply.

8. MAINTENANCE

8.1 to Refer 7515 TSM

8.5

8.6 PA Transistor Selection

The PA transistors are fitted in matched pairs to optimise amplifier performance. The gain grouping of the Motorola transistor type MRF421 is identified by a coloured dot. Only transistors of the same dot colour should be fitted.

8.7 Replacement of PA and Driver Transistors

It is neither necessary nor desirable to remove the PA assembly to replace either the driver or PA transistors. This avoids disturbing the thermal joint between the PA and the heatsink.

 $6BA \times 3/8$ inch cheese head screws are used to secure both the driver and PA transistor flanges, with the exception of the $6BA \times 5/8$ inch cheese head screw used to make the chassis connection for C20.

The leads of both driver and PA transistors are folded up at the ends to facilitiate removal with pliers.

8.7.1 Driver Transistor

- (a) Unsolder the emitter end of R12, bend the resistor clear and remove the flange fixing screws.
- (b) Using a de-soldering tool or "solder-wick" remove the bulk of the solder from each lead. Gently pull the lead away from the PCB while heating the joint. Clear away any excess solder from the emitter, base and collector pads.

Thoroughly clean the transistor mating surface on the mounting block with a cloth or tissue.

- (c) Form the leads of the replacement transistors using the discarded transistor as a guide.
- (d) Coat the transistor flange with a thin film of thermal compound (e.g. Jermyn Thermaflow A30).

8.7.1 (Cont.)

- (e) Fit the transistors (check orientation) and tighten the flange fixing screws evenly.
- (f) Carefully solder the transistor leads, this should be carried out quickly using a very hot tipped soldering iron.
- (g) Reconnect R12.
- (h) Adjust the bias current refer para 6.3.8.

8.7.2 Output Transistor

- (a) Unsolder the base ends of R16 and R17, bend them clear and remove the flange fixing screws. Bend C20 clear also.
- (b) Follow steps (b) to (f) in para 8.7.1.

to (f)

For para (f) – Silver loaded solder (e.g. "Capalloy" 62% tin, 36% lead and 2% silver) must be used to make the connection between the inner emitter leads and the ground plane. If silver loaded solder is not used the silver palladium terminations on chip capacitor C18 will dissolve. The chip capacitor may not be resoldered more than once or twice and small stocks of chip capacitors should be held if replacement of output transistors is undertaken.

(g) Reconnect R16 and R17 and ensure that C20 has been properly fitted.

(h) Readjust bias current – refer para 6.3.8.

9. PARTS LISTS - INDEX

- 9.1 Parts List (7303) as per 7515 TSM
- 9.2 Parts List (7515) as per 7515 TSM
- 9.3 Parts List (Common) as per 7515 TSM
- 9.4 Parts List (AC/DC PSU) as per 7515 TSM

- does not apply

- 9.5 Parts List (7727)
 - 9.5.1 Main Frame
 - 9.5.2 PA Assy.
 - 9.5.3 PA Filter Assy.

CCT REF	DESCRIPTION	MANUFACTURER	NOTES
9.5PART	s LIST (7727)		
9.5.1	Main Frame		
	Main PCB Assy. P.A. Assy. Filter PCB Assy 1 Filter PCB Assy 2	08-00946 08-01177 08-01179 08-01180	
	Microphone Assy Battery Cable Assy	08-00066 08-00987	
SW1 SW2	Switch Switch	MSP MSP	14-00085 14-00087
	Nylon Spacer	06-00265	
	Knob small round	06-00015	Option CAN, CBN CAW, CBW

For remainder of items in main assy refer 7515 T.S.M.

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CCT REF

MANUFACTURER NOTES

9.5.2 P.A. Assy.

All components prefixed	3
Printed circuit board	07-00401
Complete sub-assy.	08-01 171

CAPACITORS

C1	150pF	2%	100∨	N750 Ceramic	Philips 632	
C2	270pF	2%	100∨	N750 Ceramic	Philips 632	
C3	330pF	2%	100∨	N750 Ceramic	Philips 632	
C4	0.047µF	+80% -20%	50V	Ceramic	ELNA TL	
C5	68µF	+50% -10%	16V	Electrolytic	Philips 015	
C6	4.7µF	20%	25V	Tantalum	ITT TAG	
C7	39pF	2%	100V	N750 Ceramic	Philips 632	
C8	330pF	2%	100∨	N750 Ceramic	Philips 632	
C9	68µF	+50% -10%	167	Electrolytic	Philips 015	
C10	0.047µF	+80% -20%	50V	Ceramic	Elna TL	
C11	0.047µF	+80% -20%	50V	Ceramic	Elna TL	
C12	0.047µF	+80% -20%	50V	Ceramic	Elna TL	
C13	0.1µF	20%	50V	Ceramic Chip	Vitramon	
C14	180pF	2%	100V	N750 Ceramic	Philips 632	
C15	NOT USE	D				
C16	0.1µF	20%	50∨	Ceramic Chip	Vitramon	
C17	3300pF	10%	100∨	Ceramic	Philips 630	
C18	0.47µF	20%	50V	Ceramic Chip	Vitramon	
C19	68µF	+50% -10%	16V	Electrolytic	Philips 015	
C20	0.047µF	10%	100∨	Polyester	Elna Greencap	Must be an Elna Greencap 100V
C21	180pF	2%	100∨	N750 Ceramic	Philips 632	
C22	3300pF	10%	100∨	Ceramic	Philips 630	
C23	0.047µF	+80% -20%	50V	Ceramic	Elna TL	
C24	1000pF	10%	100V	Ceramic	Philips 630	
C25	lμF	20%	35V	Tantalum	ITT TAG	
C26	lΌμF	20%	25V	Tantalum	ITT TAG	
C27	0.047µF	+80% -20%	50V	Ceramic	Elna TL	
C28	10µF [']	20%	25V	Tantalum	ITT TAG	

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MANUFACTURER NOTES

9.5.2 P.A. Assy (Cont.)

RESISTORS

R1 R2 R3 R4 R5 R6 R7 R8 R9 R10 R11 R12 R13 R14 R15	15 180 10K 470 220 220K 5.6K 1K 100 10 100 1 1 1 1	5%	1/3W	Philips CR25 Philips CR25
R16 R17	6.8 6.8	5% 5%	1/2W 1/2W	Philips CR37 Philips CR37
R18	4.7	5%	1/3W	Philips CR25
R10 R19	39	5%	1/3W	Philips CR25
R20	220	5%	1/0// 1W	Philips CR68
R21	150	5%	1/3W	Philips CR25
R22	180 1K	Q / C	1/ 011	Philips CR25
R23	330			Philips CR25
R24	330			Philips CR25
R25	39			Philips CR25
R26	270			Philips CR25
R27	NOT USE	D		1
R28	NOT USE	D		
R29	NOT USE	D		
R30	NOT USE	D		
R31	NOT USE	D		
R32	220	5%	1/3W	Philips CR25
R33	68K			Philips CR25
R34	NOT USE		- /	
R35	15K	5%	1/3W	Philips CR25
R36	100K			Philips CR25

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MANUFACTURER NOTES

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.2 P.A. Assy	. (Cont.)

RESISTORS (Cont.)

R37	١K			Philips CR25	
R38	3.3K			Philips CR25	
R39	SOT			Philips CR25	Set Power output
R40	10K			Philips CR25	•
R41	1.5K			Philips CR25	
R42	820			Philips CR25	
R43	1K			Philips CR25	
R44	SOT			Philips CR25	Set Driver Bias
R45	3.3			Philips CR25	
R46	220	5%	1W	Philips CR68	
R47	100	5%	1/3W	Philips CR25	
R48	120			Philips CR25	
R49	100			Philips CR25	
R50	820			Philips CR25	
R51	SOT			Philips CR25	Set PA Bias
R52	560			Philips CR25	
R53	SOT	5%	1/2W	Philips CR37	Set Tx Lamp
R54	33	5%	1/3W	Philips CR25	•
				•	

INDUCTORS

L1	15µH Inductor	4480044	
L2	15µH Inductor	4480044	
L3	Inductor	Philips	Inductor made from
			screening beads
L4	15µH Inductor	4480044	
L5	15µH Inductor	4480044	

DIODES

D1-D5	To Spec. 23-10002
D6	NOT FITTED
D7, D8	To Spec. 23-10002

CCT REF	[DESCRIPTION	MANUFACTU	JRER NOTES
9.5.2	P.A. Assy. (Co	nt.)	######################################	n na serie de la constante de l L
	DIODES (Cont.)			
Z1 Z2 Z3 Z4	NOT USED BZX79C6V8 BZX79C5V6 IN5338B 5V1	5%	Philips Philips I.R.	
a a	TRANSFORMERS			
T1 T2 T3 T8 T9 T10	Output Transformer Input Interstage Driver Reverse Power Volto SWR Current Forward Power Volto	age	08-01178 448 0087 448 0074 448 0089 448 0097 448 0099 448 0098	
	TRANSISTORS			
TR 1 TR2 TR3 TR4 TR5 TR6 TR7	SL610C BC548 BF199 BF199 BLY87CF BLY87CF MRF421)		Plessey Philips Philips Philips Philips Philips Motorola	Philips BLY568CF may
TR8 TR9 TR10 TR11 TR12 TR13 TR14) Matched MRF421) TIP32A BC338 LM358N BD135 C/W BD675 C/W BC548	pair Mini Dip Mica Washer Mica Washer	Motorola Texas Philips National Philips Philips Philips	be substituted with other component change
- 	RELAY			
A/1	Relay 12V	1 C/O 320Ω	Siemens	V23027-A002-A101

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9.5.3 P.A. Filter Assy

9.5.3.1 P.A. Filter Assy 1.

All components prefixed	4
Printed circuit board	07-00409
Complete sub assy.	08-00839

CAPACITORS

C1 C2 C3 C4 C5	1800pF 2200pF 1000pF 6800pF 1500pF	5% 5% 5% 5% 5%	125V 125V 125V 63V 125V	Polystyrene Polystyrene Polystyrene Polystyrene Polystyrene	Philips 425) Philips 425) Philips 425) Philips 424) Philips 425)	Broadcast Station Filter
C6 C7 C8 C9 C10 C11 C12 C13 C14	270pF 39pF 100pF 100pF 150pF 100pF 47pF 39pF 27pF	5% 5% 5% 5% 5% 5% 5% 5%	500V 500V 500V 500V 500V 500V 500V 500V	Polystyrene Ceramic Polystyrene Polystyrene Polystyrene Ceramic Ceramic Ceramic	Philips 427) Philips 555) Philips 427) Philips 427) Philips 427) Philips 427) Philips 555) Philips 555) Philips 555)	7-11 MHz Filter
C15 C16 C17 C18 C19 C20 C21 C22 C23	390pF 100pF 220pF 100pF 220pF 180pF 68pF 68pF 47pF	5% 5% 5% 5% 5% 5% 5% 5%	250V 500V 500V 500V 500V 500V 500V 500V	Polystyrene Polystyrene Polystyrene Polystyrene Polystyrene Polystyrene Polystyrene Polystyrene Ceramic	Philips 426) Philips 427) Philips 427) Philips 427) Philips 427) Philips 427) Philips 427) Philips 427) Philips 555)	4.6-7 MHz Filter
	RESISTO	RS				

10K

5%

1/3W

Philips CR25

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9.5.3 P.A. Filter Assy

9.5.3.1 P. A. Filter Assy 1 (Cont.)

INDUCTORS

LI	5.6µH	4470188
L2	3.8µH	4470101
L3	Filter	4470181
L4	Filter	4470182
L5	Filter	4470183
L6	Filter	4470179
L7	Filter	4470180

MANUFACTURER NOTES

9.5.3 P.A. Filter Assy.

9.5.3.2 P. A. Filter Assy. 2

All components prefixed	8
Printed circuit board	07-00410
Complete sub assy	08-00840

CAPACITORS

C1 C2 C3 C4 C5 C6 C7 C8	390pF 330pF 270pF 220pF 330pF 270pF 220pF 68pF	5% 5% 5% 5% 5% 5% 5%	250V 250V 500V 250V 500V 500V 500V	Polystyrene Polystyrene Polystyrene Polystyrene Polystyrene Polystyrene Polystyrene Polystyrene	Philips 426) Philips 426) Philips 427) Philips 427) Philips 426) Philips 427) Philips 427) Philips 427) Philips 427)	3-4.6 MHz Filter
C9 C10 C11 C12 C13 C14 C15 C16 C17	470pF 470pF 390pF 470pF 270pF 470pF 470pF 330pF 100pF	5% 5% 5% 5% 5% 5% 5% 5%	250V 250V 250V 250V 500V 250V 250V 250V	Polystyrene Polystyrene Polystyrene Polystyrene Polystyrene Polystyrene Polystyrene Polystyrene Polystyrene	Philips 426) Philips 426) Philips 426) Philips 426) Philips 427) Philips 426) Philips 426) Philips 426) Philips 427)	2–3 MHz Filter

INDUCTORS

L1	Filter	4470177
L2	Filter	4470176
L3	Filter	4470177
L4	Filter	4470178
L5	Filter	4470174
L6	Filter	4470175

CCT REF

10. LIST OF DRAWINGS

(North Street

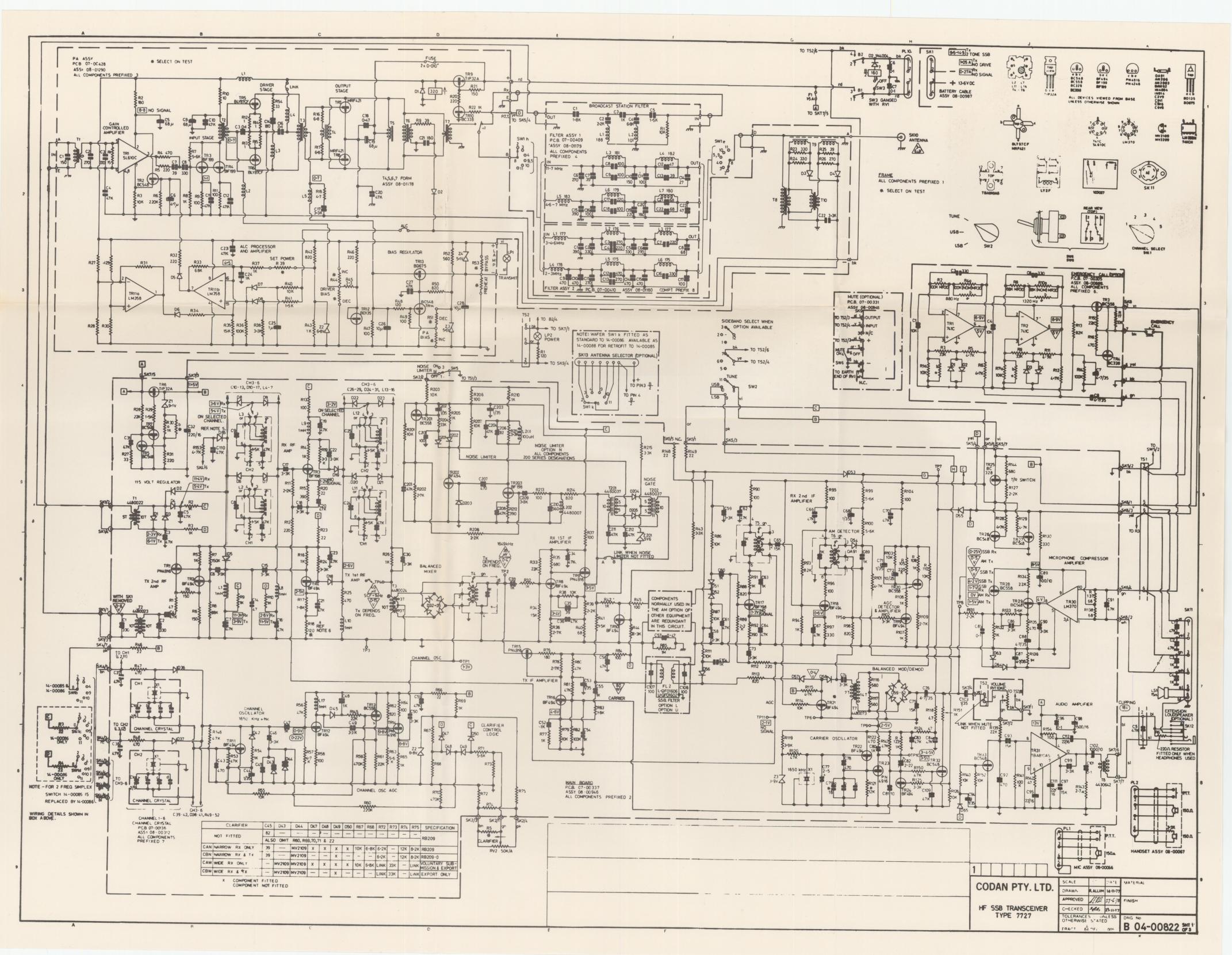
Nelses.

 Block Diagram Refer 7515 T.S.M.
 03-00181

 Circuit Diagram Type 7727
 04-00822

Component Layouts and Assembly Drawings :-

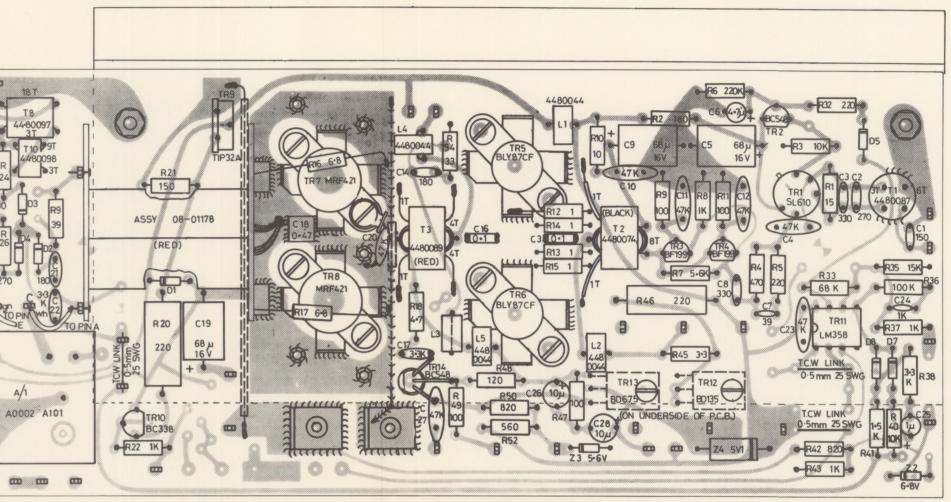
P.A. Assy	08-01177
P.A. Filter PCB Assy 1	08-01179
P.A. Filter PCB Assy 2	08-01180





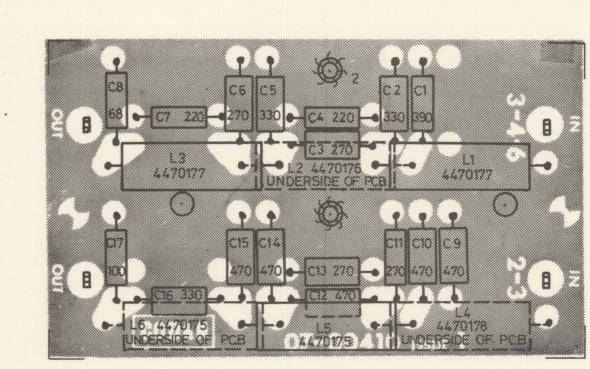
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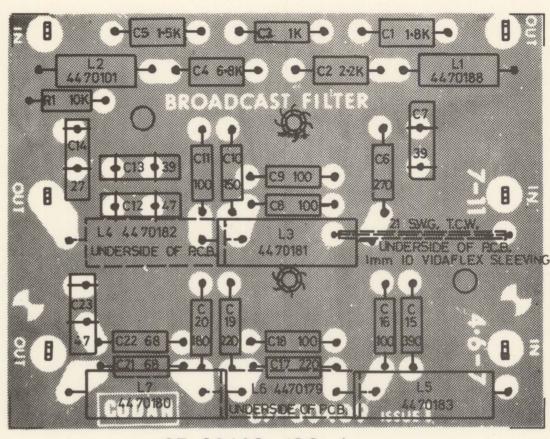
P.C.B. 07-00428 ISS 2.

P.A. ASSY 08-01290 ISS.1



07-00410 ISS 1

P.A. FILTER P.C.B. ASSY. 2 08-01180 ISS. 1



07-00409 ISS 1

P.A. FILTER P.C.B. ASSY. 1 08-01179 ISS. 1

CIRCUITNOTES

1.	Capacitor values up to 99,000 are expressed in pF , values greater are expressed in uF .
2.	

3. D control voltages.

4. * "Select on test" component.

5. Transformer colour code: or = (orange) Part Number 4470031 gn = (green) . Part Number 4470032

MEASUREMENT NOTES

 The voltages given are measured with respect to the negative rail unless otherwise shown. Voltages are typical and may vary between units. Signal voltages will vary with channel frequency.

AC voltages shown $\overbrace{\ldots,V}$ are measured with an oscilloscope probe 10M ohm and 7pF expressed in P-P units.

All measurements are made with a DC supply of 13.6 volts.

 AC voltages shown view are expressed in P-P units measured as above in the SSB Transmit mode.

> The voltages shown for the Main PCB Assy are obtained with a single tone input from an isolated source (i.e. transformer coupled – see drg. 04–00316) which produces approximately 0.9V P-P at test point TP8 (corresponds to onset of compression in the Microphone Amplifier). Remove connector SK1 for these measurements to disable the PA.

The voltages shown for the PA Assy are obtained with a two-tone input under the same conditions a above. The figures indicate the range of voltages to be expected when delivering 100 watts PEP to a 50 ohm load (corresponds to 200 P-P envelope voltage). Due to ALC action a single tone will not drive the PA to its full voltage swing capability. A single tone input will give approximately 0.8 times the voltages shown for a two-tone input.

PA VOLTAGES (Typical)

Freq.	TR7/8	TR5/6			TR 3/4	Battery
MHz	Coll.Vpp	Coll.Vpp	Base Vpp	Emitt.Vpk	Coll.Vpp	Curr. Amps
2	24	6	0.66	0.17	5.2	10.2
4	24	6	0.8	0.19	6.2	10
6	25	6.6	0.92	0.22	8	10.7
8	25	8	1.1	0.27	8.4	10.7
10	25	10.4	1.4	0.33	10.4	12.4

Note: 1) Use low capacitance probe to measure voltage at TR3/4 collectors, \leqslant 7pF.

Use an earth as close as practicable to point of measurement. Wind excess probe earth lead around probe to minimise loop area.
 ALC voltage is 7.6 V pk for two tone input.

3. Voltages shown A refer to the emf of a 50 ohm signal generator, applied at that point with respect to the negative rail, which causes the AGC voltage at test point 2TP11 to increase by 0.3 volts from its no signal value in the receive mode.

4. Frequencies shown at monitor points are for USB operation (LSB filter).

CONTROL RAIL VOLTAGES

Control Rail	Test point	SSB R×	SSB Tx	TUNE Rx	TUNE Tx
В	SK5/7	11.5	11.5	11.5	11.5
С	SK5/6	11.5	0	11.5	0
D	SK5/4	0	11.5	• 0	11.5
G	TP7	10.4	10.7	0	10.7
н	SK5/3	11.5	11.5	11.5	0

 For unused channels, link the lugs of SW1b corresponding to these channels to the channel 6 lug (SW1b/11). If all 6 channels are used omit components 2R153, 2C110 and 2D66.

6. Resistor 2R18 may be selected to set exciter gain.

